

CLAIMS

What is claimed is:

- 1 1. An equalizer comprising:
 - 2 a feedforward filter adapted to receive a first input
 - 3 signal and provide a first output signal;
 - 4 an adaptive coefficient generator adapted to receive the
 - 5 first input signal and a second signal and provide tap
 - 6 coefficients to the feedforward filter;
 - 7 a slicer adapted to receive a slicer input signal and
 - 8 provide a slicer output signal;
 - 9 a slicer timing alignment block adapted to receive the
 - 10 slicer input signal and provide a second output signal, wherein
 - 11 the slicer output signal is subtracted from the second output
 - 12 signal to generate an error signal;
 - 13 a tap timing alignment block adapted to receive the slicer
 - 14 output signal and provide a third output signal;
 - 15 a first low pass filter adapted to receive the third output
 - 16 signal and the error signal and provide a fourth output signal,
 - 17 wherein the fourth output signal is multiplied with the third
 - 18 output signal to provide a feedback signal which is added to the
 - 19 first output signal to generate the slicer input signal; and
 - 20 a second low pass filter adapted to receive the error
 - 21 signal and provide a mean square error signal.
- 1 2. The equalizer of Claim 1, further comprising a
- 2 register block adapted to receive the mean square error signal
- 3 from the second low pass filter and the tap coefficients from
- 4 the adaptive coefficient generator.

1 3. The equalizer of Claim 1, wherein a bandwidth estimate
2 is obtained for a communication channel based on correlation
3 coefficient values determined when the slicer output signal is
4 open-circuited and fixed values are provided for the tap
5 coefficients.

1 4. The equalizer of Claim 1, wherein a channel
2 identification estimation is obtained for a communication
3 channel based on subtracting a second input signal ($r(t)*h(t)$)
4 from the slicer input signal, where $r(t)$ is a random signal,
5 $h(t)$ is an unknown impulse response for the communication
6 channel, and the first input signal is statistically equivalent
7 to $r(t)$, and determining the tap coefficients corresponding to a
8 least mean square optimal set.

1 5. The equalizer of Claim 4, wherein the slicer output
2 signal is open-circuited.

1 6. The equalizer of Claim 1, wherein an optical signal-
2 to-noise ratio estimation is obtained for a communication
3 channel based on the mean square error signal, the tap
4 coefficients, and the slicer input signal or based on the tap
5 coefficients.

1 7. The equalizer of Claim 1, wherein a bit error rate
2 estimation is obtained for a communication channel based on the
3 mean square error signal and the slicer input signal.

1 8. The equalizer of Claim 1, wherein a chromatic
2 dispersion estimate is obtained by estimating a bandwidth roll-
3 off and utilizing look-up table values.

1 9. The equalizer of Claim 1, wherein a chromatic
2 dispersion estimate is obtained by computing a spectral response
3 of the feedforward filter and determining a weighted average of
4 a group delay across the frequencies to estimate a group delay
5 variation as a measure of the chromatic dispersion.

1 10. The equalizer of Claim 1, wherein a polarization mode
2 dispersion estimate is obtained by determining a frequency at
3 which a spectral response is minimal from an estimated power
4 spectral density.

1 11. The equalizer of Claim 1, wherein the equalizer is a
2 fractionally-spaced linear equalizer which provides a continuous
3 time adaptation for a communication channel.

1 12. The equalizer of Claim 1, wherein the adaptive
2 coefficient generator time-aligns the error signal with the
3 first input signal.

1 13. The equalizer of Claim 1, wherein the slicer timing
2 alignment block time-aligns the slicer input signal with the
3 slicer output signal.

1 14. The equalizer of Claim 1, wherein the tap timing
2 alignment block time-aligns the slicer output signal with a
3 symbol period.

1 15. The equalizer of Claim 1, wherein the second signal
2 comprises the error signal or the tap coefficients.

1 16. The equalizer of Claim 1, wherein the second signal
2 comprises the error signal.

1 17. An equalizer comprising:

2 means for receiving a first input signal and providing an
3 equalized output signal;

4 means for receiving the first input signal and providing
5 tap coefficients to the means for providing the equalized output
6 signal;

7 a slicer adapted to receive a slicer input signal and
8 provide a slicer output signal;

9 means for generating an error signal based on the slicer
10 output signal;

11 means for generating a feedback signal, which is summed
12 with the equalized output signal to generate the slicer input
13 signal; and

14 means for generating a mean square error signal based on
15 the error signal.

1 18. The equalizer of Claim 17, further comprising means
2 for storing the tap coefficients and the mean square error
3 signal.

1 19. The equalizer of Claim 17, wherein the equalizer is
2 employed to determine at least one of a bandwidth estimate, a
3 channel identification estimate, a signal-to-noise ratio
4 estimate, a chromatic dispersion estimate, and a polarization
5 mode dispersion estimate for a communication channel associated
6 with the equalizer.

1 20. The equalizer of Claim 17, wherein the equalizer is a
2 fractionally-spaced transversal filter with decision feedback
3 and least mean square-based adaptation to provide a continuous
4 time adaptation for a communication channel.

1 21. A method for providing a bandwidth estimate for a
2 communication channel using an equalizer, the method comprising:

3 switching off a slicer of the equalizer;

4 setting tap coefficients of a feedforward filter of the
5 equalizer to fixed values; and

6 calculating correlation coefficient values.

1 22. The method of Claim 21, wherein the correlation
2 coefficient values are calculated based on the following
3 equation, $\tilde{c}_i \approx E(p(t) \cdot p(t-i-\tau'))$, $0 \leq i \leq N$,

4 where N is the number of tap coefficients, E is the
5 expected value operator, and p is an input signal received by
6 the feedforward filter having eight multipliers.

1 23. The method of Claim 21, further comprising:

2 changing one or more of the values for the tap
3 coefficients; and

4 calculating a set of correlation coefficient values.

1 24. The method of Claim 23, further comprising calculating
2 a power spectral density based on the set of correlation
3 coefficient values.

1 25. The method of Claim 24, wherein the power spectral
2 density calculation utilizes a windowing function.

1 26. The method of Claim 24, wherein the power spectral
2 density calculation utilizes a windowing function.

1 27. The method of Claim 21, wherein the set of correlation
2 coefficient values are calculated based on the following
3 equation, $\tilde{c}_{i,j} \approx E(p(t-j \cdot \tau) \cdot p(t-i \cdot \tau'))$, $0 \leq i \leq N$, $0 \leq j \leq N$, where N is the
4 number of tap coefficients.

1 28. The method of Claim 21, further comprising changing
2 timing control ratios of the equalizer to calculate further sets
3 of correlation coefficient values.

1 29. A method for providing a channel identification
2 estimate for a communication channel using an equalizer, the
3 method comprising:

4 receiving a first input signal by a feedforward filter of
5 the equalizer, wherein the feedforward filter provides a first
6 output signal;

7 receiving a second input signal denoted as $r(t) \cdot h(t)$, where
8 $h(t)$ represents an unknown channel impulse response for the
9 communication channel and $r(t)$ represents a random signal;

10 subtracting the second input signal from the first output
11 signal to provide a difference signal; and

12 determining adaptively a set of tap coefficients for the
13 equalizer that minimizes the energy of the difference signal
14 within the equalizer.

1 30. The method of Claim 29, wherein $r(t)$ is approximately
2 statistically equivalent to the first input signal.

1 31. The method of Claim 30, wherein $r(t)$ is generated
2 using a pseudo-random binary sequence or additive white Gaussian
3 noise.

1 32. The method of Claim 29, wherein the set of tap
2 coefficients are from the feedforward filter and decision
3 feedback circuits of the equalizer.

1 33. The method of Claim 29, wherein the set of tap
2 coefficients correspond to a least mean square set of optimal
3 tap coefficients that regenerate the unknown channel.

1 34. A method for providing an optical signal-to-noise
2 ratio estimate for a communication channel using an equalizer,
3 the method comprising:

4 calculating an unbiased electrical signal-to-noise ratio
5 based on an input signal to a slicer of the equalizer and a mean
6 square error signal generated by the equalizer;

7 calculating an electrical signal-to-noise ratio based on
8 the unbiased electrical signal-to-noise ratio and tap
9 coefficients of a feedforward filter of the equalizer; and

10 calculating the optical signal-to-noise ratio based on the
11 electrical signal-to-noise ratio.

1 35. The method of Claim 34, wherein the optical signal-to-
2 noise ratio is the square root of the electrical signal-to-noise
3 ratio.

1 36. A method for providing a bit error rate estimate for a
2 communication channel using an equalizer, the method comprising:

3 calculating an unbiased electrical signal-to-noise ratio
4 based on an input signal to a slicer of the equalizer and a mean
5 square error signal generated by the equalizer; and

6 calculating the bit error rate based on the unbiased
7 electrical signal-to-noise ratio.

1 37. The method of Claim 34, wherein the bit error rate is
2 calculated using the following equation, $BER = Q(0.5 \cdot \alpha \cdot \sqrt{SNR_{e,u}})$

3 where α is a constant.

1 38. A method for providing an optical signal-to-noise
2 ratio estimate for a communication channel using an equalizer,
3 the method comprising:

4 calculating an electrical signal-to-noise ratio based on
5 tap coefficients of the equalizer; and

6 calculating the optical signal-to-noise ratio based on the
7 electrical signal-to-noise ratio.

1 39. The method of Claim 38, wherein the electrical signal-
2 to-noise ratio is calculated using the following equation,

3
$$SNR_e = \frac{(\sum_{i=0}^N c_i^2)}{1 - \frac{1}{\sum_{i=0}^N c_i + f - 1}},$$
 where f is the frequency and N is the

4 number of tap coefficients.

1 40. A method for providing a chromatic dispersion estimate
2 for a communication channel using an equalizer, the method
3 comprising:

4 determining a bandwidth roll-off within the communication
5 channel; and

6 estimating the chromatic dispersion by utilizing a look-up
7 table and the results of the bandwidth roll-off determination.

1 41. A method for providing a chromatic dispersion estimate
2 for a communication channel using an equalizer, the method
3 comprising:

4 calculating a spectral response of a feedforward filter of
5 the equalizer;

6 determining a group delay at discrete frequencies for a
7 frequency spectrum; and

8 determining a weighted average of the group delays to
9 estimate a group delay variation as a measure of the chromatic
10 dispersion.

1 42. The method of Claim 41, wherein the spectral response
2 is determined from the following equation, $P(\omega) = \sum_{i=0}^N c_i \cdot e^{j\omega\tau_i}$, where
3 N is the number of tap coefficients.

1 43. The method of Claim 41, wherein the determining of the
2 group delay includes a low end and a high end of the frequency
3 spectrum.

1 44. A method for providing a polarization mode dispersion
2 estimate for a communication channel using an equalizer, the
3 method comprising:

4 determining a frequency f_0 at which a spectral response is
5 minimal; and

6 calculating the polarization mode dispersion based on the
7 frequency f_0 .

1 45. The method of Claim 44, wherein the polarization mode
2 dispersion is calculated using the following equation,

3
$$\tau_{pmd} = \frac{1}{2 \cdot f_0}$$